

A Novel w-Transform of Bicomplex numbers and their properties

C V Rao

Department of General Studies, Jubail Industrial College, P. O. Box - 10099, Jubail Industrial City-31961, Saudi Arabia cvrao@jic.edu.sa

ABSTRACT

In this paper, I have derived a new w-Transform similar to z-transform of Bicomplex number. Also, I have derived some useful properties in W-Transform of Bicomplex numbers. I have solved w-transform of some standard sequences. I hope these results highly useful in the field of Electrical and electronics engineering fields.

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1. Introduction

Corrado Segre (year 1848) [1], [2] introduced the tessarines in a series of articles in the philosophical magazine, H. Hankel developed a general system of “complex numbers with several real generators” in 1967.

atessarine is a hypercomplex number of the form

$$w = a + ib + jc + kd, \quad a, b, c, d \in \mathbb{R}$$

$$\text{Where } i^2 = j^2 = -1, \quad ij = ji = k \text{ and } k^2 = 1$$

1.1 Basic definitions

We know [2],[3],[4],[5],[6],[7],[8] the set of complex numbers \mathbb{C} defined as follows

$\mathbb{C} = \{a + ib : a, b \in \mathbb{R}\}$ where i is an imaginary number, by using the extension of one imaginary to three imaginary numbers i, j and k such that $i^2 = j^2 = -1, ij = ji = k$ and $k^2 = 1$

Definition 1 : (Bicomplex number)

Bicomplex numbers defined as $\mathbb{T} = \{z_1 + jz_2 : z_1, z_2 \in \mathbb{C}\}$

Definition 2 : (Bicomplex conjugates) There are three possible conjugates. Let

$w \in \mathbb{T}$ and $z_1, z_2 \in \mathbb{C}$ such that $w = z_1 + jz_2$

$$w^* = \bar{z}_1 + j\bar{z}_2$$

$$w^\# = z_1 - jz_2$$

$$w' = \bar{z}_1 - j\bar{z}_2$$

Here \bar{z}_1 and \bar{z}_2 are standard complex conjugates of complex numbers \mathbb{C}

Definition 3 : (Bicomplex moduli) There are four moduli possible. Let

$w \in \mathbb{T}$ and $z_1, z_2 \in \mathbb{C}$ such that $w = z_1 + jz_2$

$$|w|_i^2 = w w^*$$

$$|w|_j^2 = w w^\#$$

$$|w|_k^2 = w w'$$

Euclidean \mathbb{R}^4 – norm defined as

$$\|w\| = \sqrt{|z_1|^2 + |z_2|^2}$$

Definition 4 : (Bicomplex multiplicative inverse)

$w \in T$ and $z_1, z_2 \in \mathbb{C}$ such that $w = z_1 + jz_2$ is invertible if

$$w^{-1} = \frac{z_1}{|z_1|^2 + |z_2|^2} + j \frac{z_2}{|z_1|^2 + |z_2|^2}$$

Definition 5: (Idempotent basis) A Bicomplex number $w = z_1 + jz_2$ has the following unique Idempotent representation

$$w = (z_1 - iz_2)e_1 + (z_1 + iz_2)e_2 \quad \text{where } z_1, z_2 \in \mathbb{C}$$

$$e_1 = \frac{1+k}{2} e_1 = \frac{1-k}{2} \quad , \quad ij = k, \quad k^2 = 1$$

$$\text{Or } w = ((a + d) + (c - d))e_1 + ((a - d) + (c + d))e_1$$

Definition 6: (Projection) Two projection operators P_1 and P_2 has the following unique Idempotent representation defined from $\mathbb{T} \rightarrow \mathbb{C}$

$$\text{as } P_1(z_1 + jz_2) = z_1 - iz_2$$

$$P_2(z_1 + jz_2) = z_1 + iz_2$$

$$\begin{aligned} \text{Then } w = z_1 + jz_2 &= P_1(w)e_1 + P_2(w)e_2 \\ &= P_1(z_1 + jz_2)e_1 + P_2(z_1 + jz_2)e_2 \end{aligned}$$

Definition 7: (Trigonometric representation of Bicomplex numbers)

$$\begin{aligned} \text{If } w = z_1 + jz_2 &= \sqrt{z_1^2 + z_2^2} \left(\frac{z_1}{\sqrt{z_1^2 + z_2^2}} + \frac{z_2}{\sqrt{z_1^2 + z_2^2}} \right) \\ &= r_c (\cos \theta_c + j \sin \theta_c) \\ &= r_c e^{j\theta_c} \end{aligned}$$

Where $r_c = \sqrt{z_1^2 + z_2^2} = P_1(w)P_2(w)$,

$$\cos \theta_c = \frac{z_1}{\sqrt{z_1^2 + z_2^2}} \text{ and } \sin \theta_c = \frac{z_2}{\sqrt{z_1^2 + z_2^2}} \quad (1)$$

$$\text{Also from (1)} \frac{z_1}{z_2} = \frac{\cos \theta_c}{\sin \theta_c} \Rightarrow \frac{z_1 - iz_2}{z_1 + iz_2} = \frac{(\cos \theta_c - i \sin \theta_c)}{(\cos \theta_c + i \sin \theta_c)}$$

$$\Rightarrow \frac{P_1(w)}{P_2(w)} = \frac{(\cos \theta_c - i \sin \theta_c)}{(\cos \theta_c + i \sin \theta_c)}$$

$$\Rightarrow \frac{P_1(w)}{P_2(w)} = \frac{e^{-i\theta_c}}{e^{i\theta_c}}$$

$$\Rightarrow \frac{P_1(w)}{P_2(w)} = e^{-i2\theta_c}$$

Take ln on both sides

$$\Rightarrow \theta_c = i \ln \sqrt{\frac{P_1(w)}{P_2(w)}}$$

Z-Transforms:

Definition : [9] , [10] Given a sequence x_n then Z-transform defined as

$X(z) = \sum_{n=0}^{\infty} x_n z^{-n}$, z is a complex number

This transformation produces a new representation of x_n denoted by $X(z)$.

From $X(z)$ we can get original sequence x_n (inverse z -Transform) by finding n th power of the coefficient of z^{-1} .

2. w-Transform:

Definition : Given a sequence x_n then w -Transform defined as

$X(w) = \sum_{n=0}^{\infty} x_n w^{-n}$, $w = z_1 + jz_2$ is a Bicomplex number

This transformation produces a new representation of x_n denoted by $X(w)$.

From $X(w)$ we can get original sequence x_n (inverse w -Transform) by finding n th power of the coefficient of w^{-1} .

$$X(w) = \sum_{n=0}^{\infty} x_n w^{-n}$$

or

$$\Rightarrow X(w) = \sum_{n=0}^{\infty} x_n (u^{-n} e_1 + v^{-n} e_2)$$

$$u = z_1 - iz_2 \quad v = z_1 + iz_2 \quad , z_1, z_2 \in \mathbb{C}$$

$$\Rightarrow X(w) = e_1 \sum_{n=0}^{\infty} x_n u^{-n} + e_2 \sum_{n=0}^{\infty} x_n v^{-n}$$

Example 1 : $x_n = \alpha^n$

Solution: $X(w) = \sum_{n=0}^{\infty} x_n w^{-n}$

$$\Rightarrow X(w) = \sum_{n=0}^{\infty} \alpha^n w^{-n}$$

$$\Rightarrow X(w) = \frac{w}{w - \alpha}$$

$$\Rightarrow X(w) = \frac{1}{1 - \alpha w^{-1}}$$

or

$$\Rightarrow X(w) = \sum_{n=0}^{\infty} \alpha^n (u^{-n} e_1 + v^{-n} e_2) \quad , \quad u, v \in \mathbb{C}$$

$$\Rightarrow X(w) = e_1 \sum_{n=0}^{\infty} \alpha^n u^{-n} + e_2 \sum_{n=0}^{\infty} \alpha^n v^{-n}$$

$$\Rightarrow X(w) = e_1 \sum_{n=0}^{\infty} \alpha^n u^{-n} + e_2 \sum_{n=0}^{\infty} \alpha^n v^{-n}$$

$$\begin{aligned} \Rightarrow X(w) &= e_1 \frac{u}{u-\alpha} + e_2 \frac{v}{v-\alpha} \\ \Rightarrow X(w) &= e_1 \frac{1}{1-\alpha u^{-1}} + e_2 \frac{1}{1-\alpha v^{-1}} \\ \Rightarrow X(w) &= e_1 u_1 + e_2 v_1 = w_1 \end{aligned}$$

Where $u_1 = \frac{1}{1-\alpha u^{-1}}$, $u = z_1 - iz_2$

$v_1 = \frac{1}{1-\alpha v^{-1}}$, $v = z_1 + iz_2$ and $w = z_1 + jz_2$

Note : If $\alpha = 1$ then $x_n = 1$

$$\begin{aligned} X(w) &= \sum_{n=0}^{\infty} x_n w^{-n} \\ X(w) &= \sum_{n=0}^{\infty} 1 w^{-n} \\ \Rightarrow X(w) &= \frac{1}{1-w^{-1}} \end{aligned}$$

Or $X(w) = e_1 \frac{1}{1-u^{-1}} + e_2 \frac{1}{1-v^{-1}}$, $u, v \in \mathbb{C}$
 $\Rightarrow X(w) = e_1 u_2 + e_2 v_2 = w_2$

Where $u_2 = \frac{1}{1-u^{-1}}$, $u = z_1 - iz_2$
 $v_2 = \frac{1}{1-v^{-1}}$, $v = z_1 + iz_2$ and $w = z_1 + jz_2$

Example 2 : $x_n = n$

Solution: $X(w) = \sum_{n=0}^{\infty} x_n w^{-n}$

$$\begin{aligned} \Rightarrow X(w) &= \sum_{n=0}^{\infty} n w^{-n} \\ \Rightarrow X(w) &= \frac{w}{(w-1)^2} \\ \Rightarrow X(w) &= \frac{1}{w(1-w^{-1})^2} \end{aligned}$$

Or

$$\Rightarrow X(w) = e_1 \frac{u}{(u-1)^2} + e_2 \frac{v}{(v-1)^2}, \quad u, v \in \mathbb{C}$$

Example 3. Dirac delta sequence: $x_n = \delta(n - n_0)$

Solution: $X(w) = \sum_{n=0}^{\infty} x_n w^{-n}$

$$\begin{aligned} \Rightarrow X(w) &= \sum_{n=0}^{\infty} \delta(n - n_0) w^{-n} \\ \Rightarrow X(w) &= w^{-n_0} \end{aligned}$$

$$\Rightarrow X(w) = (w^{-1})^{n_0}$$

or

$$\Rightarrow X(w) = \left(\frac{w^*}{|w|^2_i}\right)^{n_0} = \left(\frac{w^\#}{|w|^2_j}\right)^{n_0} = \left(\frac{w^*}{|w|^2_k}\right)^{n_0}$$

Put $n_0 = 1$ we get the impulse response of the unit delay

$$\begin{aligned} X(w) &= \sum_{n=0}^{\infty} \delta(n-1) w^{-n} \\ &\Rightarrow X(w) = w^{-1} \end{aligned}$$

or

$$\Rightarrow X(w) = \frac{w^*}{|w|^2_i} = \frac{w^\#}{|w|^2_j} = \frac{w^*}{|w|^2_k}$$

Example 4. Exponential function : $x_n = e^{\alpha n}$, $\alpha \in \mathbb{R}$

Solution: $X(w) = \sum_{n=0}^{\infty} x_n w^{-n}$

$$\begin{aligned} \Rightarrow X(w) &= \sum_{n=0}^{\infty} e^{\alpha n} w^{-n} \\ \Rightarrow X(w) &= \sum_{n=0}^{\infty} (e^{\alpha} w^{-1})^n \\ \Rightarrow X(w) &= (1 - e^{\alpha} w^{-1})^{-1} \\ \Rightarrow X(w) &= \frac{w}{w - e^{\alpha}} \\ \Rightarrow X(w) &= \frac{1}{1 - e^{\alpha} w^{-1}} \end{aligned}$$

3.Properties of w-Transform:

1. Linearity property:

$$a_1 x_n + a_2 y_n \xleftrightarrow{w} w a_1 X(w) + a_2 Y(w)$$

Proof: $X(w) = \sum_{n=0}^{\infty} (a_1 x_n + a_2 y_n) w^{-n}$

$$= \sum_{n=0}^{\infty} (a_1 x_n w^{-n} + a_2 y_n w^{-n})$$

$$\begin{aligned}
 &= \sum_{n=0}^{\infty} a_1 x_n w^{-n} + \sum_{n=0}^{\infty} a_2 y_n w^{-n} \\
 &= a_1 \sum_{n=0}^{\infty} x_n w^{-n} + a_2 \sum_{n=0}^{\infty} y_n w^{-n} \\
 &= a_1 X(w) + a_2 Y(w)
 \end{aligned}$$

2. The time delay property:

$$\begin{aligned}
 &x_{n-n_0} \xleftrightarrow{w} w^{-n_0} X(w) \\
 &\text{and } x_{n-1} \xleftrightarrow{w} w^{-1} X(w) \\
 \text{proof: } X(w) &= \sum_{n=0}^{\infty} x_{n-n_0} w^{-n} \\
 &\text{replace } n - n_0 \text{ my } m
 \end{aligned}$$

$$\begin{aligned}
 X(w) &= \sum_{m=0}^{\infty} x_m w^{-(m+n_0)} \\
 X(w) &= w^{-n_0} \sum_{m=0}^{\infty} x_m w^{-m} \\
 X(w) &= w^{-n_0} X(w)
 \end{aligned}$$

Special case if $n_0 = 1$

$$X(w) = w^{-1} X(w)$$

Thus $x_{n-n_0} \xleftrightarrow{w} w^{-n_0} X(w)$

$$\text{and } x_{n-1} \xleftrightarrow{w} w^{-1} X(w)$$

3. convolution and w-Transform:

$$x_n * h_n \xleftrightarrow{w} H(w) X(w)$$

Proof : Let $y_n = x_n * h_n$

Take w-transform on both sides

$$y_n = \sum_{l=0}^{\infty} h_l x_{n-l}$$

Take the w-transform on both sides using superposition and general delay property

$$Y(w) = \sum_{l=0}^{\infty} h_l w^{-l} X(w)$$

$$\Rightarrow Y(w) = \left(\sum_{l=0}^{\infty} h_l w^{-l} \right) X(w)$$

$$\Rightarrow Y(w) = H(w)X(w)$$

hence

$$x_n * h_n \xleftrightarrow{W} H(w)X(w)$$

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